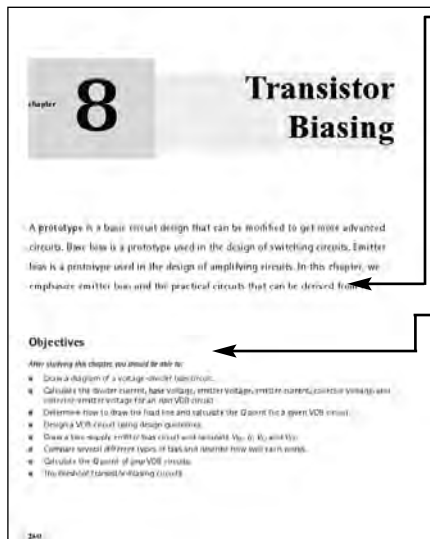


Guided Tour

Learning Features

Many new learning features have been incorporated into the seventh edition of *Electronic Principles*. These learning features, found throughout the chapters, include:



CHAPTER INTRODUCTION

Each chapter begins with a brief introduction setting the stage for what the student is about to learn.

CHAPTER OBJECTIVES

Chapter Objectives provide a concise statement of expected learning outcomes.

VOCABULARY

A comprehensive list of new vocabulary words alerts the students to key words found in the chapter. Within the chapter, these key words are highlighted in bold print the first time used.



PRACTICE PROBLEMS

Students can obtain critical feedback by performing the Practice Problems that immediately follow most Examples. Answers to these problems are found at the end of each chapter.

Here is an important point: The calculations do not depend on changes in the transistor, or the temperature. This is why the Q point of this circuit is stable, almost rock-solid.

PRACTICE PROBLEM 8-1 Change the power supply voltage of Fig. 8-3 for V_{CC} .

GOOD TO KNOW

Good To Know statements, found in the margins, provide interesting added insights to topics being presented.

GOOD TO KNOW

Since $I_E \approx I_C$, Eq. 8-10 can also be shown as

$$V_{CE} = V_{CC} - I_C R_C - I_E R_E$$

or

$$V_{CE} \approx V_{CC} - I_C (R_C + R_E)$$

5. Calculate the collector-to-ground voltage by subtracting the voltage across the collector resistor from the collector supply voltage.

6. Calculate the collector-emitter voltage by subtracting the emitter voltage from the collector voltage.

Since these six steps are logical, they should be easy to remember. After you analyze a few VBE circuits, the process becomes automatic.

Example 8-1

What is the collector-emitter voltage in Fig. 8-2?

SOLUTION The voltage divider produces an unloaded output voltage of:

$$V_{AB} = \frac{2.2 \text{ k}\Omega}{10 \text{ k}\Omega + 2.2 \text{ k}\Omega} 10 \text{ V} = 1.8 \text{ V}$$

Subtract 0.7 V from this to get:

$$V_E = 1.8 \text{ V} - 0.7 \text{ V} = 1.1 \text{ V}$$

The emitter current is:

$$I_E = \frac{1.1 \text{ V}}{1 \text{ k}\Omega} = 1.1 \text{ mA}$$

Since the collector current almost equals the emitter current, we can calculate the collector-to-ground voltage like this:

EXAMPLES

Each chapter contains worked-out Examples that demonstrate important concepts or circuit operation, including circuit analysis, applications, troubleshooting, and basic design.

MULTISIM

Students can “bring to life” many of the circuits found in each chapter. A CD containing MultiSim files is included with the textbook; with these files students can change the value of circuit components and instantly see the effects, using realistic Tektronix and Agilent simulation instruments. Troubleshooting skills can be developed by inserting circuit faults and making circuit measurements. Students new to computer simulation software will find a MultiSim Primer in the appendix.

Example 8-1

What is the collector-emitter voltage in Fig. 8-2?

SOLUTION The voltage divider produces an unloaded output voltage of:

$$V_{AB} = \frac{2.2 \text{ k}\Omega}{10 \text{ k}\Omega + 2.2 \text{ k}\Omega} 10 \text{ V} = 1.8 \text{ V}$$

Subtract 0.7 V from this to get:

$$V_E = 1.8 \text{ V} - 0.7 \text{ V} = 1.1 \text{ V}$$

The emitter current is:

$$I_E = \frac{1.1 \text{ V}}{1 \text{ k}\Omega} = 1.1 \text{ mA}$$

Since the collector current almost equals the emitter current, we can calculate the collector-to-ground voltage like this:

$$V_{CE} = 10 \text{ V} - (1.1 \text{ mA})(3.6 \text{ k}\Omega) = 6.04 \text{ V}$$

The collector-emitter voltage is:

$$V_{CE} = 6.04 \text{ V} - 1.1 \text{ V} = 4.94 \text{ V}$$

Here is an important point: The calculations in this preliminary analysis do not depend on changes in the transistor, the collector current, or the temperature. This is why the Q point of this circuit is stable, almost rock-solid.

PRACTICE PROBLEM 8-1 Change the power supply voltage of Fig. 8-2 from 10 V to 15 V and solve for V_{CE} .

Example 8-2

Discuss the significance of Fig. 8-3, which shows a MultiSim analysis of the same circuit analyzed in the preceding example.

SOLUTION This really drives the point home. Here we have an almost identical answer using a computer to analyze the circuit. As you can see, the voltmeter reads 6.03 V (rounded to 2 places). Compare this to 6.04 V in the preceding example, and you can see the point. A simplified analysis has produced essentially the same result as a computer analysis.

You can expect this kind of close agreement whenever a VBE circuit has been well designed. After all, the whole point of VBE is to set the emitter bias so virtually eliminate the effects of changing the transistor, collector current, or temperature.

Transistor Biasing 243

DATA SHEETS

Full and partial component data sheets are provided for many semiconductor devices; key specifications are examined and explained. Complete data sheets of these devices can be found on the Internet.

Figure 3-16 Data sheet for 1N4001–1N4007 diodes.

1N4001–1N4007

FEATURES

- Low forward voltage drop.
- High surge current capability.

General Purpose Rectifiers

Absolute Maximum Ratings* $T_C = 25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Value	Units
V_{RRM}	Peak Reverse Voltage	50	V
I_{FSM}	Forward Surge Current	100	A
I_{TSM}	Non-repetitive Peak Forward Surge Current	100	A
I_{TSM}	Peak Storage-ridge Shock Current	30	A
T_J	Storage Temperature Range	-65 to +175	$^\circ\text{C}$
T_C	Operating Junction Temperature	-55 to +175	$^\circ\text{C}$

*The surge current capability is dependent on the thermal conductivity of the semiconductor device being mounted.

Thermal Characteristics

Symbol	Parameter	Value	Units
$R_{\theta JC}$	Power Dissipation Junction-to-Case	30	$^\circ\text{C}/\text{W}$
$R_{\theta JA}$	Thermal Resistance Junction-to-Air	50	$^\circ\text{C}/\text{W}$

Electrical Characteristics $T_C = 25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Device	Units
V_F	Forward Voltage @ 1.0 A	1N4001 1N4002 1N4003 1N4004 1N4005 1N4006 1N4007	V
V_{RM}	Maximum Peak Reverse Current	100	μA
I_{FSM}	Forward Surge Current @ 100 μs	100	A
C_T	Total Capacitance	100	pF
C_J	Junction Capacitance	50	pF
f_T	Transition Frequency @ $V_F = 0.5 \text{ V}$, $I_F = 1.0 \text{ mA}$	50	MHz

Figure 7-22 Phototransistor. (a) Open base gives maximum sensitivity. (b) Variable base resistor changes sensitivity. (c) Typical phototransistor.

$I_{C20} = \beta I_B$
 where I_B is the reverse minority-carrier current. This says that the collector current is higher than the original reverse current by a factor of β .
 The collector diode is sensitive to light as well as heat. In a phototransistor, light passes through a window and strikes the collector-base junction. As the light increases, I_B increases, and so does I_{C20} .

Phototransistor versus Photodiode
 The main difference between a phototransistor and a photodiode is the current gain β . The same amount of light striking both devices produces β times more current in a phototransistor than in a photodiode. The increased sensitivity of a phototransistor is a big advantage over that of a photodiode.
 Figure 7-22a shows the schematic symbol of a phototransistor. Notice the open base. This is the usual way to operate a phototransistor. You can control the sensitivity with a variable base return resistor (Fig. 7-22b), but the base is usually left open to get maximum sensitivity to light.
 The price paid for increased sensitivity is reduced speed. A phototransistor is more sensitive than a photodiode, but it cannot turn on and off as fast. A photodiode has typical output currents in microamperes and can switch on and off in nanoseconds. The phototransistor has typical output currents in milliamperes but switches on and off in microseconds. A typical phototransistor is shown in Fig. 7-22c.

Optocoupler
 Figure 7-23a shows an LED driving a phototransistor. This is a much more sensitive optocoupler than the LED photodiode discussed earlier. The idea is straightforward. Any changes in V_i produce changes in the LED current, which changes the current through the phototransistor. In turn, this produces a changing voltage across the collector-emitter terminals. Therefore, a signal voltage is coupled from the input circuit to the output circuit.
 Again, the big advantage of an optocoupler is the electrical isolation between the input and output circuits. Stated another way, the common for the input circuit is different from the common for the output circuit. Because of this, no conductive path exists between the two circuits. This means that you can ground one

Figure 7-23 (a) Optocoupler with LED and phototransistor. (b) Optocoupler IC.

COMPONENT PHOTOS

Photos of actual electronic devices bring students closer to the device being studied.

SUMMARY TABLES

Summary Tables have been included at important points within many chapters. Students use these tables as an excellent review of important topics, and as a convenient information resource.

Summary Table 8-1		Main Bias Circuits		
Type	Circuit	Calculations	Characteristics	Where used
Base bias		$I_B = \frac{V_{BB} - 0.7 \text{ V}}{R_B}$ $I_C = \beta I_B$ $V_{CE} = V_{CC} - I_C R_C$	Few parts, β dependent, fixed base current.	Switch, digital
Emitter bias		$V_{BE} = V_{BB} - 0.7 \text{ V}$ $I_B = \frac{V_{BE}}{R_B}$ $V_{CE} = V_{CC} - I_C R_C$ $V_{CE} = V_{CE} - V_{CE}$	Fixed emitter current, β independent	I_C driver, amplifier

Figure 7-15 npn transistor.

Out-of-Circuit Tests
 A transistor is commonly tested using a DMM set to the diode test range. Figure 7-15 shows how an npn transistor resembles two back-to-back diodes. Each pn junction can be tested for normal forward and reverse biased readings. The collector to emitter can also be tested and should result in an overrange indication with either DMM polarity connection. Since a transistor has three leads, there are six DMM polarity connections possible. These are shown in Fig. 7-16a. Notice that only two polarity connections result in approximately a 0.7 V reading. Also important to note here is that the base lead is the only connection common to both 0.7 V readings and it requires a (+) polarity connection. This is also shown in Fig. 7-16b.
 A pnp transistor can be tested using the same technique. As shown in Fig. 7-17, the pnp transistor also resembles two back-to-back diodes. Again, using the DMM in the diode test range, Fig. 7-18a and 7-18b show the results for a normal transistor.
 Many DMMs have a special β_{DC} or I_{DC} test function. By placing the transistor's leads into the proper slots, the forward current gain is displayed. This current gain is for a specified base current or collector current and V_{CE} . You can check the DMM's manual for the specific test condition.
 Another way to test transistors is with an ohmmeter. You can begin by measuring the resistance between the collector and the emitter. This should be very high in both directions because the collector and emitter diodes are back to back in series. One of the most common troubles is a collector-emitter short, pro-

Figure 7-16 NPN DMM Readings. (a) Polarity connections; (b) pn junction readings.

+	Reading
B E	0.7
E B	OL
B C	0.7
C B	OL
C E	OL
E C	OL

COMPONENT TESTING

Students will find clear descriptions of how to test individual electronic components using common equipment such as digital multimeters (DMMs).

