Monday, June 27, 2011

CHAPTER 13

P.P. 13.1 For mesh 1,

$$141.42 + j141.42 = 4(1 + j2)\mathbf{I_1} + j\mathbf{I_2}$$
(1)

For mesh 2,

- $0 = j\mathbf{I}_1 + (10 + j5)\mathbf{I}_2$ (2)
- For the matrix form $\begin{bmatrix} 141.42 + j141.42 \\ 0 \end{bmatrix} = \begin{bmatrix} 4 + j8 & j \\ j & 10 + j5 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$ $\Delta = j100, \ \Delta_2 = 141.42 j141.42$ $\mathbf{I_2} = \Delta_2/\Delta = (141.42 j141.42)/j100$ $\mathbf{V_0} = 10\mathbf{I_2} = 10(-1.4142 j1.4142) = \mathbf{20}\angle -\mathbf{135^{\circ} V}$

P.P. 13.2 Since I_1 enters the coil with reactance 2Ω and I_2 enters the coil with reactance 6Ω , the mutual voltage is positive. Hence, for mesh 1,

$$100 \angle 60^{\circ} = (5 + j2 + j6 - j \ 3x2)I_1 - j6I_2 + j3I_2$$

or
$$100 \angle 60^{\circ} = (5 + j2)I_1 - j3I_2$$
(1)
For mesh 2,
$$0 = (j6 - j4)I_2 - j6I_1 + j3I_1$$

or
$$I_2 = 1.5I_1$$
(2)
Substituting this into (1),
$$100 \angle 60^{\circ} = (5 - j2.5)I_1$$

$$I_1 = (100 \angle 60^{\circ})/(5.59 \angle -26.57^{\circ}) = 17.889 \angle 86.57^{\circ} A$$

$$I_2 = 1.5I_1 = 26.83 \angle 86.57^{\circ} A$$

P.P. 13.3 The coupling coefficient is, $k = m/\sqrt{L_1L_2} = 1/\sqrt{2x1} = 0.7071$
To obtain the energy stored, we first obtain the frequency-domain circuit shown below.

 $100\cos(\omega t)$ becomes $100 \angle 0^\circ$, $\omega = 2$



P.P. 13.4
$$Z_{in} = 4 + j8 + [3^2/(j10 - j6 + 6 + j4)]$$

= $4 + j8 + 9/(6 + j8)$
= $8.58 \angle 58.05^{\circ} \Omega$

The current from the voltage is,

$$I = V/Z = 40 \angle 0^{\circ}/8.58 \angle 58.05^{\circ} = 4.662 \angle -58.05^{\circ} A$$

P.P. 13.5

$$L_1 = 10, L_2 = 4, M = 2$$

 $L_1L_2 - M^2 = 40 - 4 = 36$
 $L_A = (L_1L_2 - M^2)/(L_2 - M) = 36/(4 - 2) = 18 \text{ H}$
 $L_B = (L_1L_2 - M^2)/(L_1 - M) = 36/(10 - 2) = 4.5 \text{ H}$
 $L_C = (L_1L_2 - M^2)/M = 36/2 = 18 \text{ H}$

Hence, we get the π equivalent circuit as shown below.



P.P. 13.6 If we reverse the direction of I_2 so that we replace I_2 by $-I_2$, we have the circuit shown in Figure (a).



We now replace the coupled coil by the T-equivalent circuit and assume $\omega = 1$.

$$L_a = 5 - 3 = 2 H$$

 $L_b = 6 - 3 = 3 H$
 $L_c = 3 H$

Hence the equivalent circuit is shown in Figure (b). We apply mesh analysis.



$$12 = i_{1}(-j4 + j2 + j3) + j3i_{2}$$

or $12 = ji_{1} + j3i_{2}$ (1)
Loop 2 produces, $0 = j3i_{1} + (j3 + j3 + 12)i_{2}$
or $i_{1} = (-2 + j4)i_{2}$ (2)
Substituting (2) into (1), $12 = (-4 + j)i_{2}$, which leads to $i_{2} = 12/(-4 + j)$
 $I_{2} = -i_{2} = 12/(4 - j) = 2.91\angle 14.04^{\circ} A$
 $I_{1} = i_{1} = (-2 + j4)i_{2} = 12(2 - j4)/(4 - j) = 13\angle -49.4^{\circ} A$

P.P. 13.7

(a)	$n = V_2/V_1 = 110/2200 = 1/20$ (a step-down transformer)
(b)	$S = V_1I_1 = 2200x5 = 11 \text{ kVA}$

(c)
$$I_2 = I_1/n = 5/(1/20) = 100 \text{ A}$$

P.P. 13.8 The 16 - j24-ohm impedance can be reflected to the primary resulting in

$$Z_{in} = 2 + (16 - j24)/16 = 3 - j1.5$$

 $I_1 \ = \ 240/(3-j1.5) \ = 240/(3.354 \angle -26.57^\circ) = \ 71.56 \angle 26.57^\circ$

$$I_2 = -I_1/n = -17.89 \angle 26.57^\circ$$
$$V_o = -j24i_2 = (24 \angle -90^\circ)(-17.89 \angle 26.57^\circ) = 429.4 \angle 116.57^\circ V$$
$$S_1 = V_1I_1 = (240)(71.56 \angle 26.57^\circ) = 17.174 \angle -26.57^\circ kVA.$$



P.P. 13.10 We should note that the current and voltage of each winding of the autotransformer in Figure (b) are the same for the two-winding transformer in Figure (a).



For the two-winding transformer,

$$S_1 = 120x0.5 = 60 \text{ VA}$$

 $S_2 = 6(10) = 60 \text{ VA}$

For the autotransformer,

$$S_1 = 120(6.5) = 780 \text{ VA}$$

 $S_2 = 130(6) = 780 \text{ VA}$

P.P. 13.11 $(I_2)^* = S_2/V_2 = 16,000/1000 = 16 \text{ A}$ Since $S_1 = V_1(I_1)^* = V_2(I_2)^* = S_2$, $V_2/V_1 = I_1/I_2$, 1000/2500 = $I_1/32$, or $I_1 = 1000 \times 16/2500 = 6.4 \text{ A}$. At the top, KCL produces $I_1 + I_0 = I_2$, or $I_0 = I_2 - I_1 = 16 - 6.4 = 9.6 \text{ A}$.

P.P. 13.12

(a)
$$S_T = (\sqrt{3})V_L I_L$$
, but $S_T = P_T / \cos\theta = 40 \times 10^6 / 0.85 = 47.0588 \text{ MVA}$

$$I_{LS} = S_T / (\sqrt{3}) V_{LS} = 47.0588 \times 10^6 / [(\sqrt{3}) 12.5 \times 10^3] = 2.174 \text{ kA}$$

(b)
$$V_{LS} = 12.5 \text{ kV}, V_{LP} = 625 \text{ kV}, n = V_{LS}/V_{LP} = 12.5/625 = 0.02$$

(c)
$$I_{LP} = nI_{LS} = 0.02x2173.6 = 43.47 \text{ A}$$

or
$$I_{LP} = S_T / [(\sqrt{3})v_{LP}] = 47.0588 \times 10^6 / [(\sqrt{3})625 \times 10^3] = 43.47 \text{ A}$$

(d) The load carried by each transformer is
$$(1/3)S_T = 15.69$$
 MVA

P.P. 13.13 The process is essentially the same as in Example 13.13. We are given the coupling coefficient, k = 0.4, and can determine the operating frequency from the value of $\omega = 4$ which implies that $f = 4/(2\pi) = 0.6366$ Hz.



$i_o = 2.012\cos(4t + 68.52^{\circ}) A$

P.P. 13.14 Following the same basic steps in Example 13.14, we first assume $\omega = 1$. This then leads to following determination of values for the inductor and the capacitor.

 $j15 = j\omega L$ leads to L = 15 H

$$-j16 = 1/(\omega C)$$
 leads to C = 62.5 mF

The schematic is shown below.



Note, if we divide V_2 by V_1 we get $1.5046 \angle -.09^\circ$ which is in good agreement that the transformer is ideal with a voltage ratio of 1:1.5 (or 2:3)!

P.P. 13.15 $V_2/V_1 = 120/13,200 = 1/110 = 1/n$

P.P. 13.16



As in Example 13.16, $n^2 = Z_L/Z_1 = 400/(2.5 \times 10^3) = 4/25$, n = 0.4

By voltage division, $V_1 = V_s/2$ (since $Z_1 = Z_L/n^2$), therefore $V_1 = 60/2 = 30$ volts, and

$$V_2 = nV_1 = (0.4)(30) = 12$$
 volts

P.P. 13.17

- (a) S = 12x60 + 350 + 4,500 = 5.57 kW
- (b) $I_P = S/V_P 5570/2400 = 2.321 \text{ A}$