

CHAPTER 9

P.P.9.1 amplitude = **30**
 phase = **-75°**
 angular frequency (ω) = 4π = **12.57 rad/s**
 period (T) = $\frac{2\pi}{\omega}$ = **0.5 s**
 frequency (f) = $\frac{1}{T}$ = **2 Hz**

P.P.9.2 $i_1 = -4 \sin(\omega t + 55^\circ) = 4 \cos(\omega t + 55^\circ + 90^\circ)$
 $i_1 = 4 \cos(\omega t + 145^\circ)$, $\omega = 377 \text{ rad/s}$

Compare this with

$$i_2 = 5 \cos(\omega t - 65^\circ)$$

indicates that the phase angle between i_1 and i_2 is

$$145^\circ + 65^\circ = 210^\circ$$

Thus, **i_1 leads i_2 by 210°**

P.P.9.3 (a) $(5 + j2)(-1 + j4) = -5 + j20 - j2 - 8 = -13 + j18$
 $5 \angle 60^\circ = 2.5 + j4.33$
 $(5 + j2)(-1 + j4) - 5 \angle 60^\circ = -15.5 + j13.67$
 $[(5 + j2)(-1 + j4) - 5 \angle 60^\circ]^* = -15.5 - j13.67 = 20.67 \angle 221.41^\circ$

(b) $3 \angle 40^\circ = 2.298 + j1.928$
 $10 + j5 + 3 \angle 40^\circ = 12.298 + j6.928 = 14.115 \angle 29.39^\circ$
 $-3 + j4 = 5 \angle 126.87^\circ$
 $\frac{10 + j5 + 3 \angle 40^\circ}{-3 + j4} = \frac{14.115 \angle 29.39^\circ}{5 \angle 126.87^\circ} = 2.823 \angle -97.48^\circ$
 $2.823 \angle -97.48^\circ = -0.3675 - j2.8$
 $10 \angle 30^\circ = 8.66 + j5$
 $\frac{10 + j5 + 3 \angle 40^\circ}{-3 + j4} + 10 \angle 30^\circ + j5 = 8.293 + j7.2$

P.P.9.4 (a) $v = 7 \cos(2t + 40^\circ)$

The phasor form is

$$\mathbf{V} = 7 \angle 40^\circ \text{ V}$$

(b) Since $-\sin(A) = \cos(A + 90^\circ)$,

$$i = -4 \sin(10t + 10^\circ) = 4 \cos(10t + 10^\circ + 90^\circ)$$

$$i = 4 \cos(10t + 100^\circ)$$

The phasor form is

$$\mathbf{I} = 4 \angle 100^\circ \text{ A}$$

P.P.9.5

(a) Since $-1 = 1 \angle \pm 180^\circ$ (we can use either sign)

$$\mathbf{V} = -25 \angle 40^\circ = 25 \angle (40^\circ - 180^\circ) = 25 \angle -140^\circ$$

The sinusoid is

$$v(t) = 25 \cos(\omega t - 140^\circ) \text{ V or } 25 \cos(\omega t + 220^\circ) \text{ V}$$

(b) $\mathbf{I} = j(12 - j5) = 5 + j12 = 13 \angle 67.38^\circ$

The sinusoid is

$$i(t) = 13 \cos(\omega t + 67.38^\circ) \text{ A}$$

P.P.9.6

$$\begin{aligned} \text{Let } v(t) &= -10 \sin(\omega t - 30^\circ) + 20 \cos(\omega t + 45^\circ) \\ &= 10 \cos(\omega t - 30^\circ + 90^\circ) + 20 \cos(\omega t + 45^\circ) \end{aligned}$$

Taking the phasor of each term

$$\mathbf{V} = 10 \angle 60^\circ + 20 \angle 45^\circ$$

$$\mathbf{V} = 5 + j8.66 + 14.142 + j14.142$$

$$\mathbf{V} = 19.142 + j22.8 = 29.77 \angle 49.98^\circ$$

Converting \mathbf{V} to the time domain

$$v(t) = 29.77 \cos(\omega t + 49.98^\circ) \text{ V}$$

P.P.9.7

Given that

$$2 \frac{dv}{dt} + 5v + 10 \int v dt = 50 \cos(5t - 30^\circ)$$

we take the phasor of each term to get

$$2j\omega \mathbf{V} + 5 \mathbf{V} + \frac{10}{j\omega} \mathbf{V} = 50 \angle -30^\circ, \quad \omega = 5$$

$$\mathbf{V} [j10 + 5 - j(10/5)] = \mathbf{V} (5 + j8) = 50 \angle -30^\circ$$

$$\mathbf{V} = \frac{50 \angle -30^\circ}{5 + j8} = \frac{50 \angle -30^\circ}{9.434 \angle 58^\circ}$$

$$\mathbf{V} = 5.3 \angle -88^\circ$$

Converting \mathbf{V} to the time domain

$$v(t) = 5.3 \cos(5t - 88^\circ) \text{ V}$$

P.P.9.8

For the capacitor,

$$\mathbf{V} = \mathbf{I} / (j\omega C), \quad \text{where } \mathbf{V} = 10 \angle 30^\circ, \quad \omega = 100$$

$$\mathbf{I} = j\omega C \mathbf{V} = (j100)(50 \times 10^{-6})(10 \angle 30^\circ)$$

$$\mathbf{I} = 50 \angle 120^\circ \text{ mA}$$

$$i(t) = 50 \cos(100t + 120^\circ) \text{ mA}$$

P.P.9.9

$$\mathbf{V}_s = 20 \angle 30^\circ, \quad \omega = 10$$

$$\mathbf{Z} = 4 + j\omega L = 4 + j2$$

$$\mathbf{I} = \mathbf{V}_s / \mathbf{Z} = \frac{20\angle 30^\circ}{4 + j2} = \frac{20\angle 30^\circ(4 - j2)}{16 + 4} = 4.472\angle 3.43^\circ$$

$$\mathbf{V} = j\omega L \mathbf{I} = j2 \mathbf{I} = (2\angle 90^\circ)(4.472\angle 3.43^\circ) = 8.944\angle 93.43^\circ$$

Therefore, $v(t) = 8.944 \sin(10t + 93.43^\circ) \text{ V}$
 $i(t) = 4.472 \sin(10t + 3.43^\circ) \text{ A}$

P.P.9.10

Let \mathbf{Z}_1 = impedance of the 1-mF capacitor in series with the 100- Ω resistor

\mathbf{Z}_2 = impedance of the 1-mF capacitor

\mathbf{Z}_3 = impedance of the 8-H inductor in series with the 200- Ω resistor

$$\mathbf{Z}_1 = 100 + \frac{1}{j\omega C} = 100 + \frac{1}{j(10)(1 \times 10^{-3})} = 80 - j100$$

$$\mathbf{Z}_2 = \frac{1}{j\omega C} = \frac{1}{j(10)(1 \times 10^{-3})} = -j100$$

$$\mathbf{Z}_3 = 200 + j\omega L = 200 + j(10)(8) = 200 + j80$$

$$\mathbf{Z}_{in} = \mathbf{Z}_1 + \mathbf{Z}_2 \parallel \mathbf{Z}_3 = \mathbf{Z}_1 + \mathbf{Z}_2 \mathbf{Z}_3 / (\mathbf{Z}_2 + \mathbf{Z}_3)$$

$$\mathbf{Z}_{in} = 100 - j100 + \frac{-j100 \times (200 + j80)}{-j100 + 200 + j80}$$

$$\mathbf{Z}_{in} = 100 - j100 + 49.52 - j95.04$$

$$\mathbf{Z}_{in} = [149.52 - j195] \Omega$$

P.P.9.11 In the frequency domain,

the voltage source is $\mathbf{V}_s = 20\angle 100^\circ$

the 0.5-H inductor is $j\omega L = j(10)(0.5) = j5$

the $\frac{1}{20}$ -F capacitor is $\frac{1}{j\omega C} = \frac{1}{j(10)(1/20)} = -j2$

Let \mathbf{Z}_1 = impedance of the 0.5-H inductor in parallel with the 10- Ω resistor

and \mathbf{Z}_2 = impedance of the (1/20)-F capacitor

$$\mathbf{Z}_1 = 10 \parallel j5 = \frac{(10)(j5)}{10 + j5} = 2 + j4 \quad \text{and} \quad \mathbf{Z}_2 = -j2$$

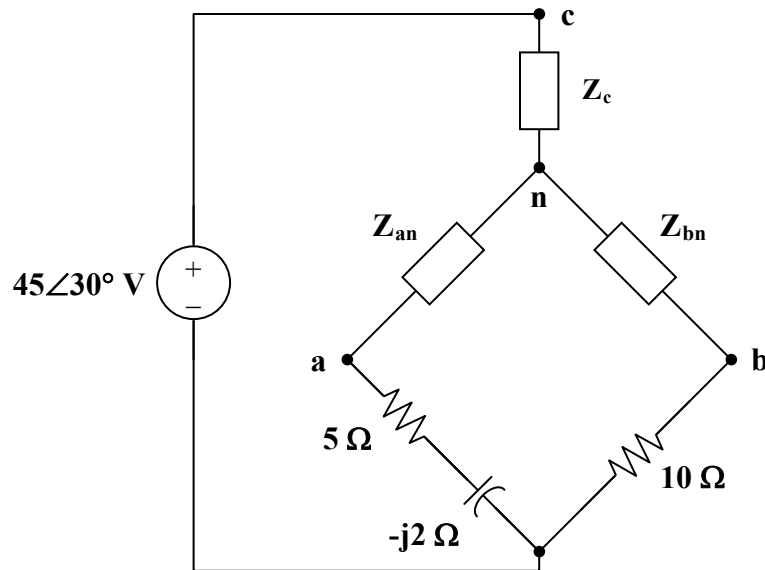
$$\mathbf{V}_o = \mathbf{Z}_2 / (\mathbf{Z}_1 + \mathbf{Z}_2) \mathbf{V}_s$$

$$\mathbf{V}_o = \frac{-j2}{2 + j4 - j2} (50\angle 30^\circ) = \frac{-j(50\angle 30^\circ)}{1 + j} = \frac{50\angle (30^\circ - 90^\circ)}{\sqrt{2}\angle 45^\circ}$$

$$\mathbf{V}_o = 35.36\angle -105^\circ$$

$$v_o(t) = 35.36 \cos(10t - 105^\circ) \text{ V}$$

P.P.9.12 We need to find the equivalent impedance via a delta-to-wye transformation as shown below.



$$\mathbf{Z}_{an} = \frac{j4(8 + j5)}{j4 + 8 + j5 - j3} = \frac{4(-5 + j8)}{8 + j6} = 0.32 + j3.76$$

$$\mathbf{Z}_{bn} = \frac{-j3(8 + j5)}{8 + j6} = \frac{3(5 - j8)(8 - j6)}{100} = -0.24 - j2.82$$

$$\mathbf{Z}_{cn} = \frac{j4(-j3)}{8 + j6} = \frac{12(8 - j6)}{100} = 0.96 - j0.72$$

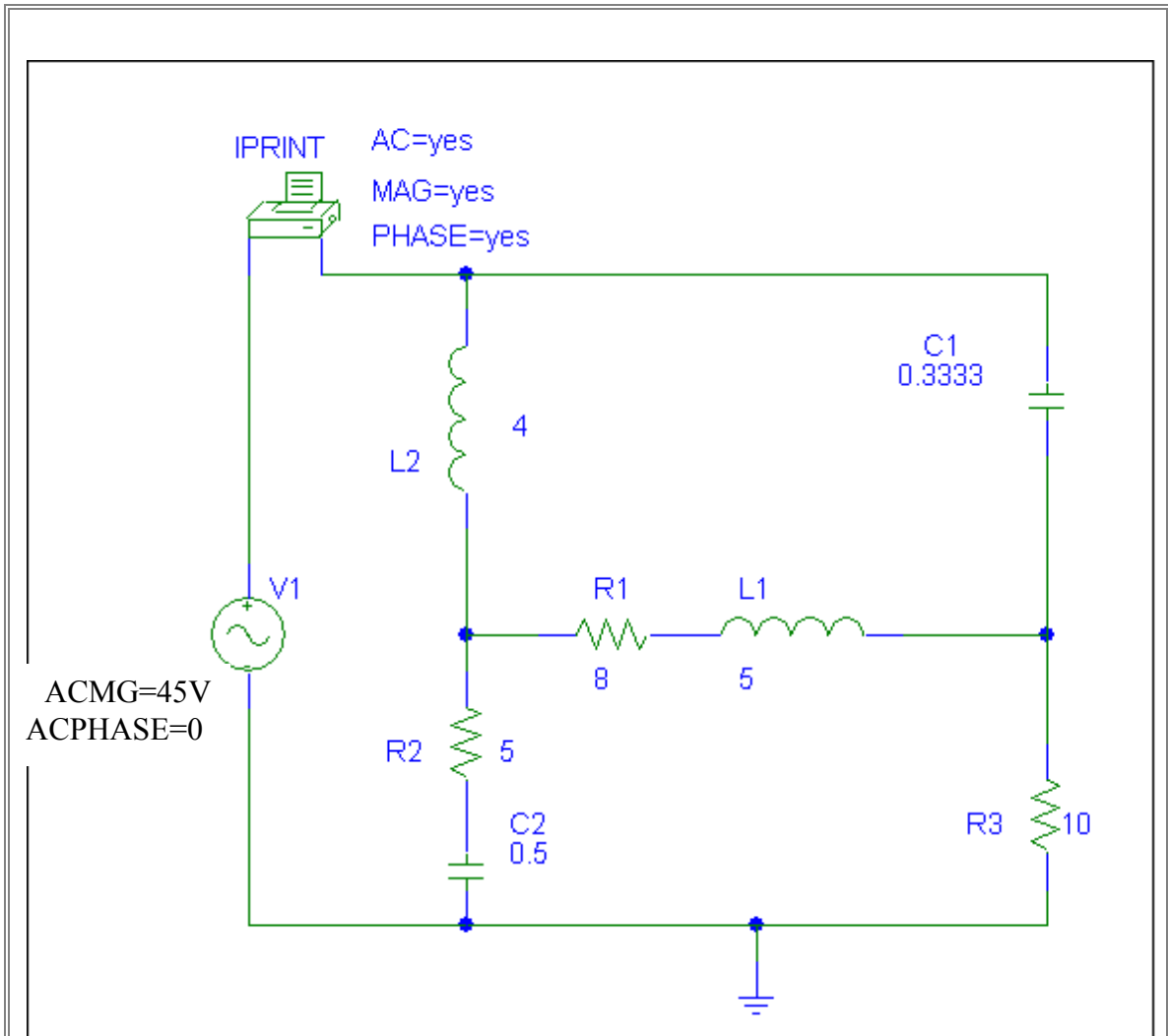
The total impedance from the source terminals is

$$\begin{aligned} \mathbf{Z} &= \mathbf{Z}_{cn} + (\mathbf{Z}_{an} + 5 - j2) \parallel (\mathbf{Z}_{bn} + 10) \\ \mathbf{Z} &= \mathbf{Z}_{cn} + (5.32 + j1.76) \parallel (9.76 - j2.82) \\ \mathbf{Z} &= \mathbf{Z}_{cn} + \frac{(5.32 + j1.76)(9.76 - j2.82)}{(5.32 + j1.76) + (9.76 - j2.82)} \\ \mathbf{Z} &= 0.96 - j0.72 + 3.744 + j0.4074 \\ \mathbf{Z} &= 4.704 - j0.3126 = 4.714 \angle -3.802^\circ \end{aligned}$$

Therefore,

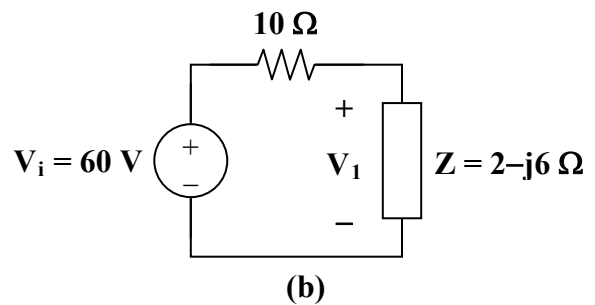
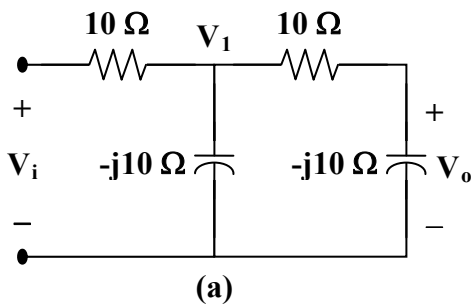
$$\begin{aligned} \mathbf{I} &= \mathbf{V} / \mathbf{Z} = \frac{45 \angle 30^\circ}{4.714 \angle -3.802^\circ} \\ \mathbf{I} &= \mathbf{9.546} \angle \mathbf{33.8^\circ} \mathbf{A} \end{aligned}$$

Let us now check this using PSpice. The solution produces the magnitude of $\mathbf{I} = 9.946\text{E}+00$, and the phase angle = $33.803\text{E}+00$, which agrees with the above answer.



P.P.9.13 To show that the circuit in Fig. (a) meets the requirement, consider the equivalent circuit in Fig. (b).

$$\mathbf{Z} = -j10 \parallel (10 - j10) = \frac{-j10(10 - j10)}{10 - j20} = \frac{-j(10 - j10)}{1 - j2} = 2 - j6 \Omega$$



$$\mathbf{V}_1 = \frac{2 - j6}{10 + 2 - j6} (60) = \frac{60}{3} (1 - j)$$

$$\mathbf{V}_o = \frac{-j10}{10 - j10} \mathbf{V}_1 = \left(\frac{-j}{1 - j} \right) \left(\frac{60}{3} \right) (1 - j) = -j20$$

$$\mathbf{V}_o = 20 \angle -90^\circ$$

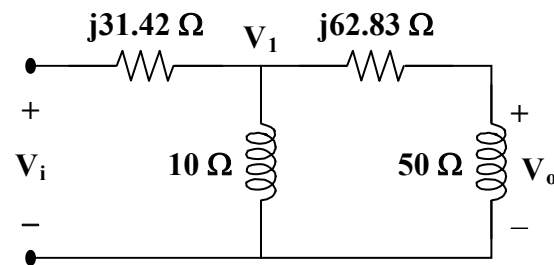
This implies that the RC circuit provides a 90° lagging phase shift.
The output voltage is = **20 V**

P.P.9.14

the 1-mH inductor is $j\omega L = j(2\pi)(5 \times 10^3)(1 \times 10^{-3}) = j31.42$

the 2-mH inductor is $j\omega L = j(2\pi)(5 \times 10^3)(2 \times 10^{-3}) = j62.83$

Consider the circuit shown below.



$$\mathbf{Z} = 10 \parallel (50 + j62.83) = \frac{(10)(50 + j62.83)}{60 + j62.83}$$

$$\mathbf{Z} = 9.205 + j0.833 = 9.243 \angle 5.17^\circ$$

$$\mathbf{V}_1 = \mathbf{Z} / (\mathbf{Z} + j31.42) \mathbf{V}_i = \frac{9.243 \angle 5.17^\circ}{9.205 + j32.253} (10)$$

$$= [(9.243 \angle 5.17^\circ) / (33.54 \angle 74.07^\circ)] 10 = 2.756 \angle -68.9^\circ$$

$$\mathbf{V}_o = \frac{50}{50 + j62.83} \mathbf{V}_1 = \frac{50(2.756 \angle -68.9^\circ)}{80.297 \angle 51.49^\circ} = 1.7161 \angle -120.39^\circ$$

Therefore,

magnitude = **1.7161 V**

phase = **120.39°**

phase shift is **lagging**

P.P.9.15 $Z_x = (Z_3 / Z_1) Z_2$

$$Z_3 = 12 \text{ k}\Omega$$

$$Z_1 = 4.8 \text{ k}\Omega$$

$$Z_2 = 10 + j\omega L = 10 + j(2\pi)(6 \times 10^6)(0.25 \times 10^{-6}) = 10 + j9.425$$

$$Z_x = \frac{12\text{k}}{4.8\text{k}}(10 + j9.425) = 25 + j23.5625 \Omega$$

$$R_x = 25, \quad X_x = 23.5625 = \omega L_x$$

$$L_x = \frac{X_x}{2\pi f} = \frac{23.5625}{2\pi(6 \times 10^6)} = 0.625 \mu\text{H}$$

i.e. a **25- Ω** resistor in series with a **0.625- μH** inductor.